

Sparse Superposition Codes for the Gaussian Channel

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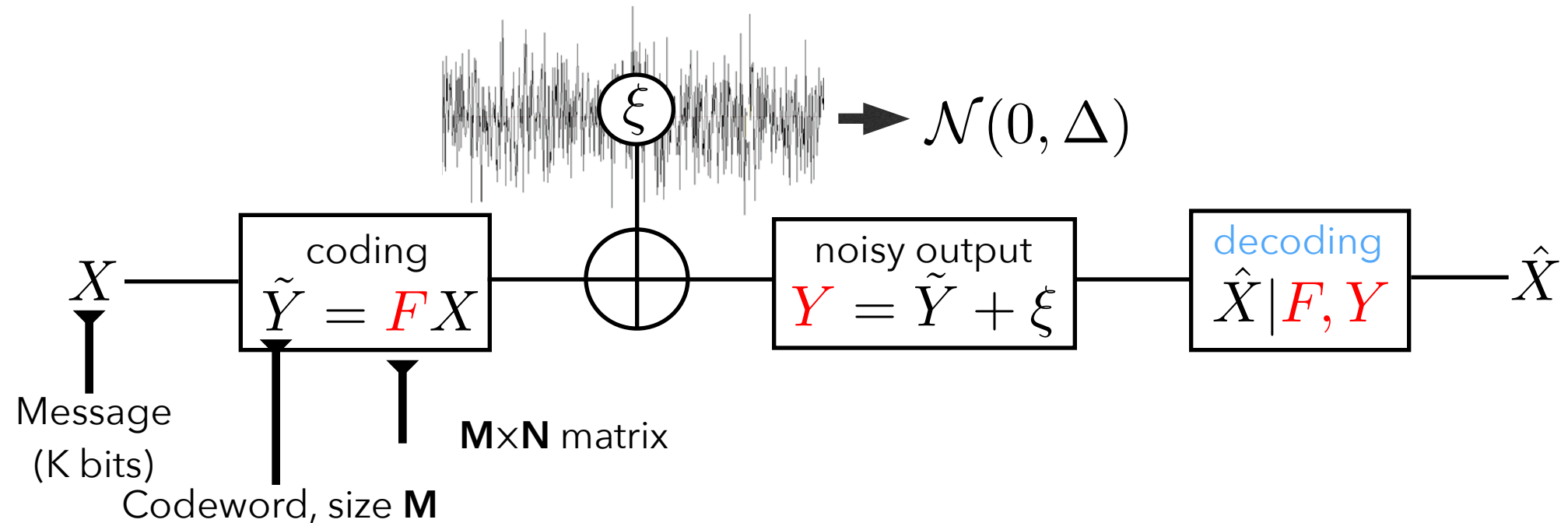


J. Barbier (ENS)

arXiv:1403.8024 presented at ISIT '14

Long version in preparation

Communication through the Gaussian Channel



Forney & Ungerboeck '98 Review
(modulation, shaping and coding)

Richardson & Urbanke '08
(LDPC and turbo codes)

Arikan '09, Abbe & Barron '11
(Polar codes)

Characteristics:

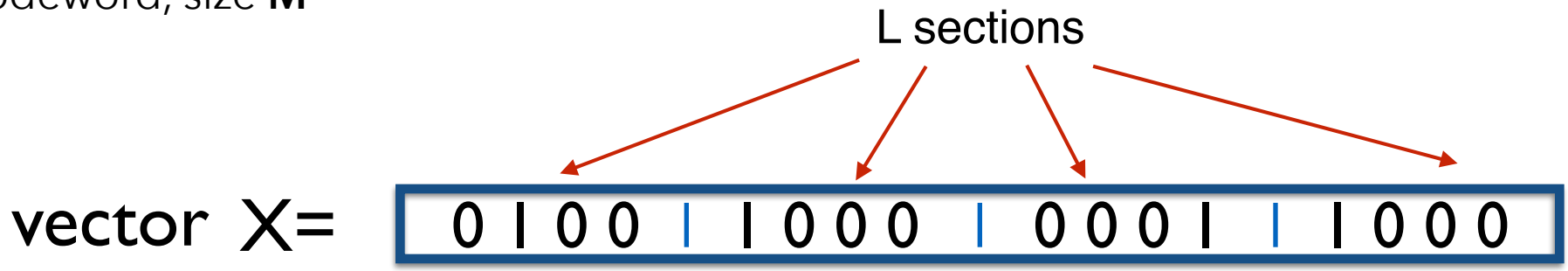
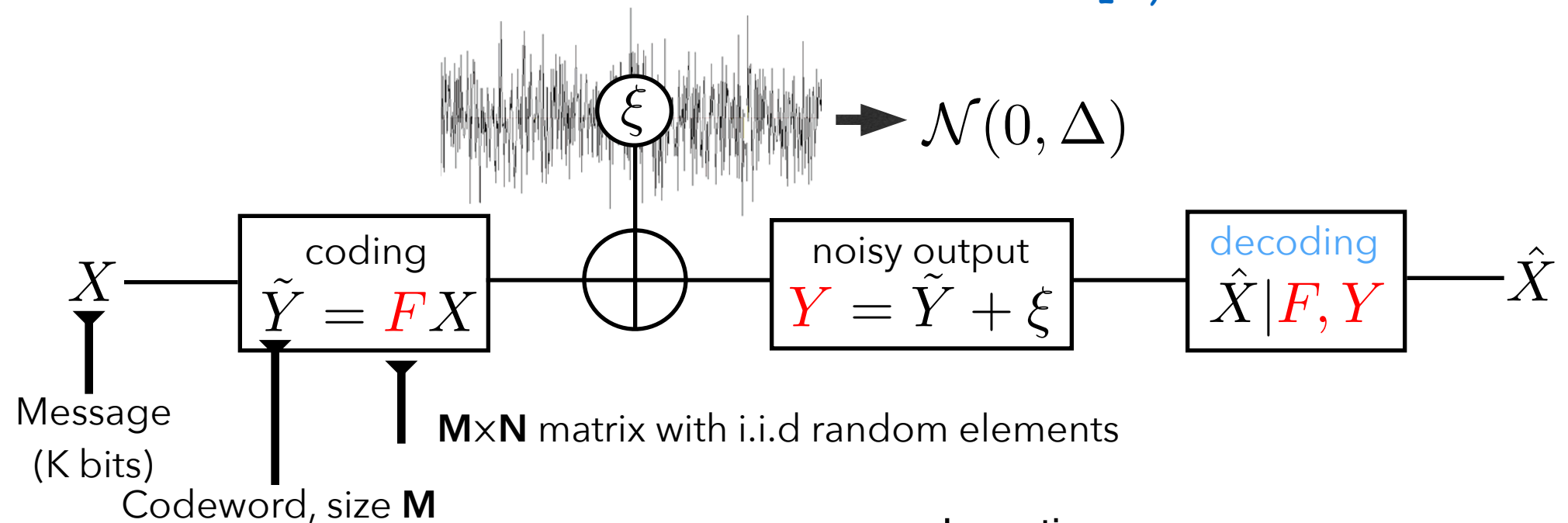
Power constraint: $\langle \tilde{Y}^2 \rangle = 1$

Signal to noise ratio: $\text{SNR} = \frac{1}{\Delta}$

Shannon Capacity: $C = \frac{1}{2} \log_2 (1 + \text{SNR})$

Sparse Superposition codes

Joseph, Barron ISIT '10



Dimension $N = LB$
 $K = L \log_2(B)$ bits of information
Rate: $R = \frac{N \log_2(B)}{BM}$

Sparse Superposition codes

Joseph, Barron ISIT '10

- For large B , the maximum likelihood solution is capacity achieving

$$\min_{\hat{x}} \|Y - F\hat{x}\|_{\ell_2} \text{ such that } \hat{x} \text{ has a single 1 per section}$$

Hard computational problem...

Sparse Superposition codes

Joseph, Barron ISIT '10

- For large B , the maximum likelihood solution is capacity achieving
- With proper power allocation, the “*adaptive successive decoder*” of Joseph and Barron achieve capacity when $B \rightarrow \infty$

vector $X =$

0	1	0	0		1	0	0	0		0	0	0		1	1	0	0	0
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Sparse Superposition codes

Joseph, Barron ISIT '10

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vector $X =$

0	c_1	0	0		c_2	0	0	0		0	0	0	c_3		c_4	0	0	0
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Power constraint: $\langle c_i^2 \rangle = 1$ $c_i^2 \propto e^{-\frac{2Ci}{L}}$

Sparse Superposition codes

Joseph, Barron ISIT '10

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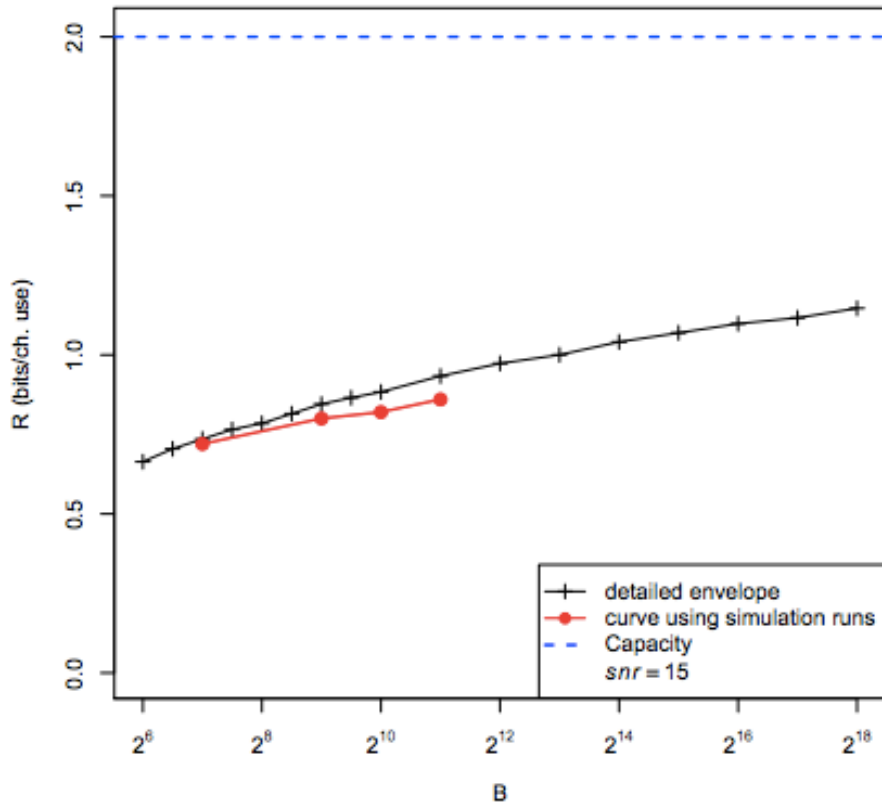
Power constraint: $\langle c_i^2 \rangle = 1$ $c_i^2 \propto e^{-\frac{2Ci}{L}}$

- In practice, however, results are FAR from capacity when B is finite

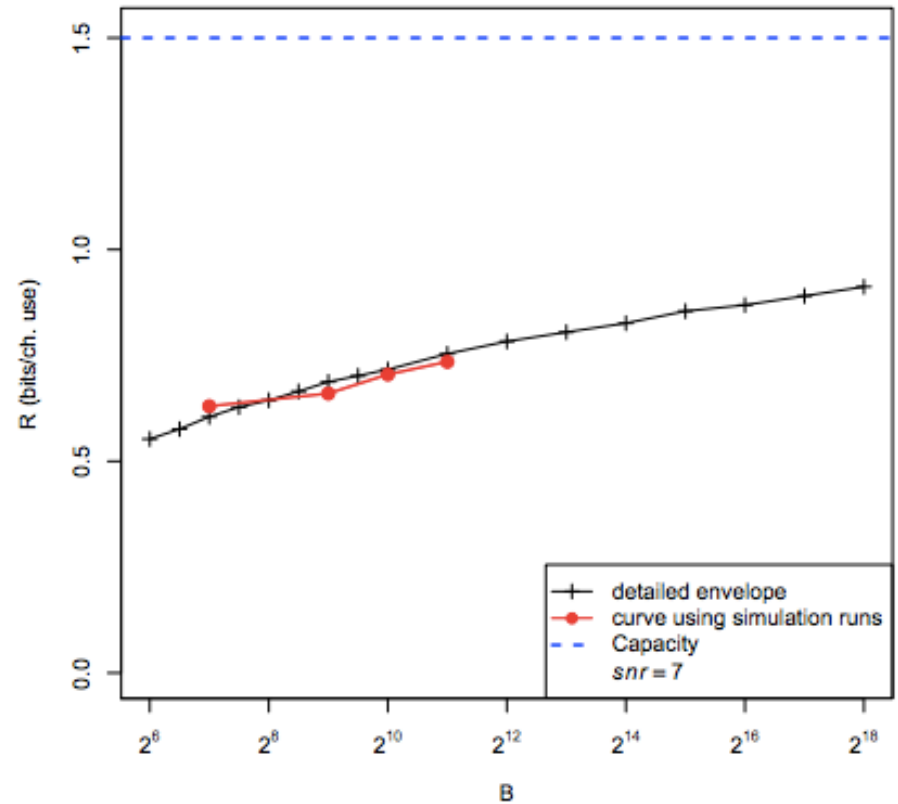
Sparse Superposition codes

Joseph, Barron ISIT '10

SNR=15



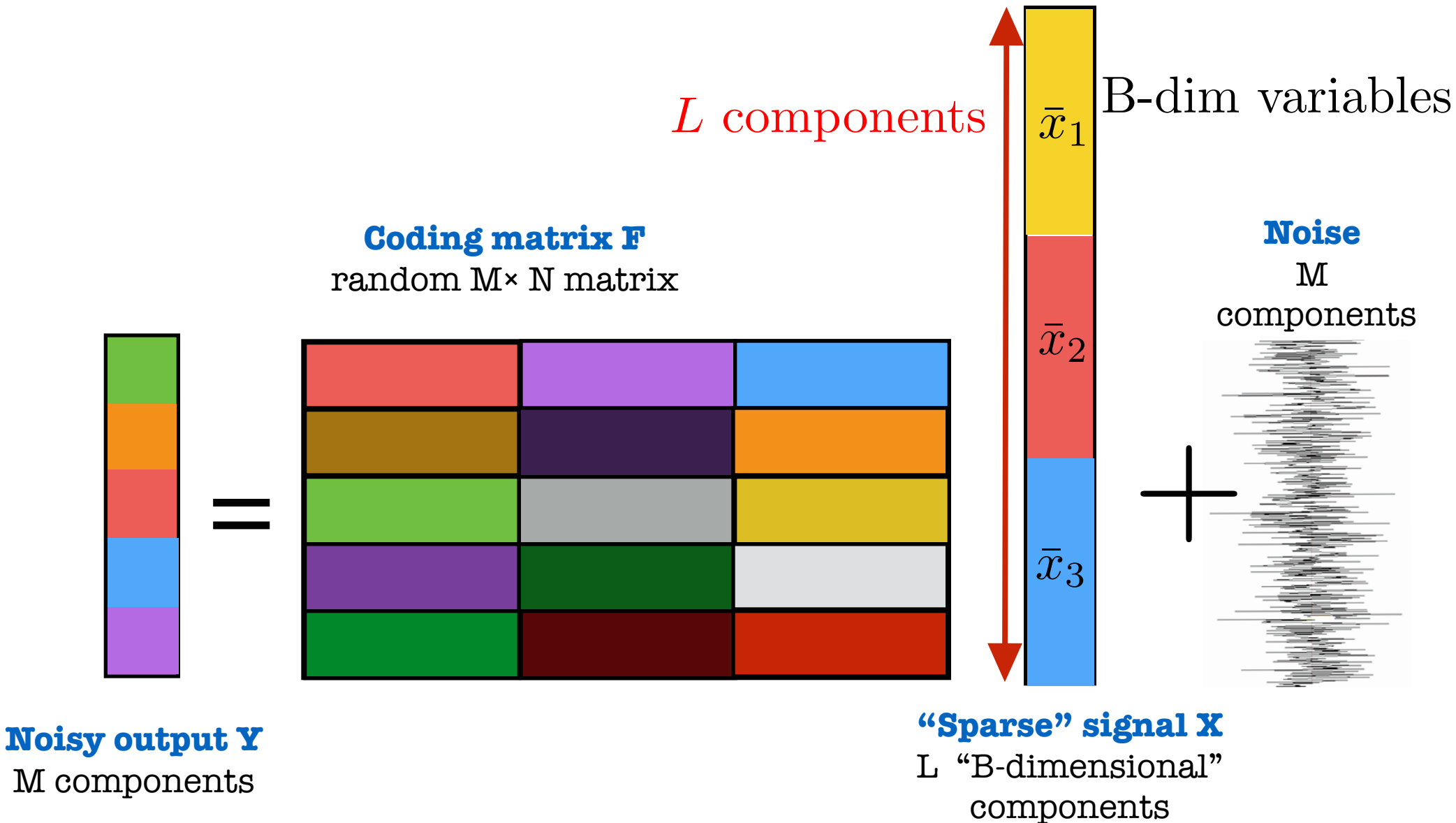
SNR=7





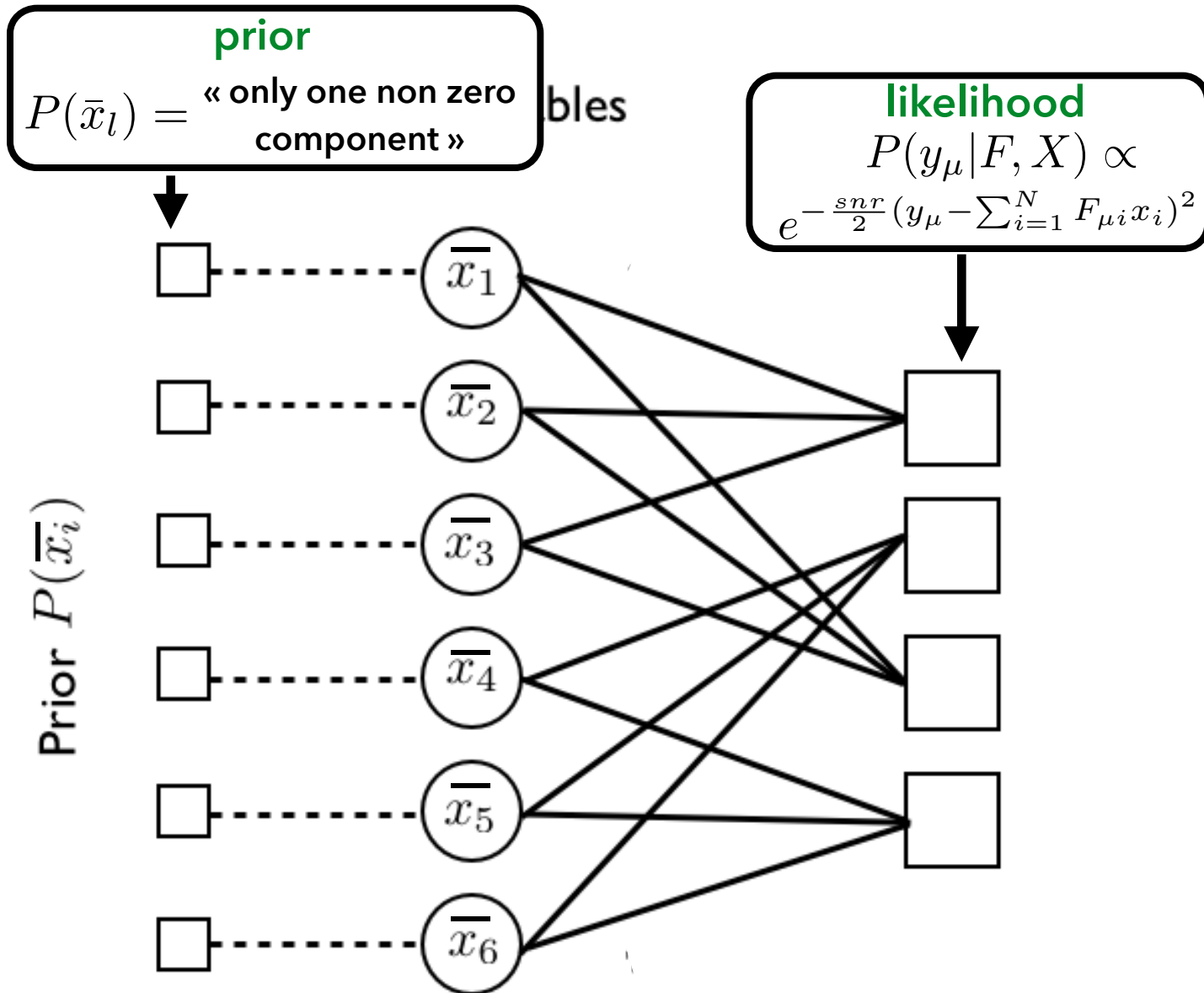
The approximate message-passing algorithm

Sparse Superposition codes = Multi-dimensional estimation problem



Graphical model

Compute the marginal of $P(X|F, Y) \propto P(X)P(Y|F, X)$



Graphical model

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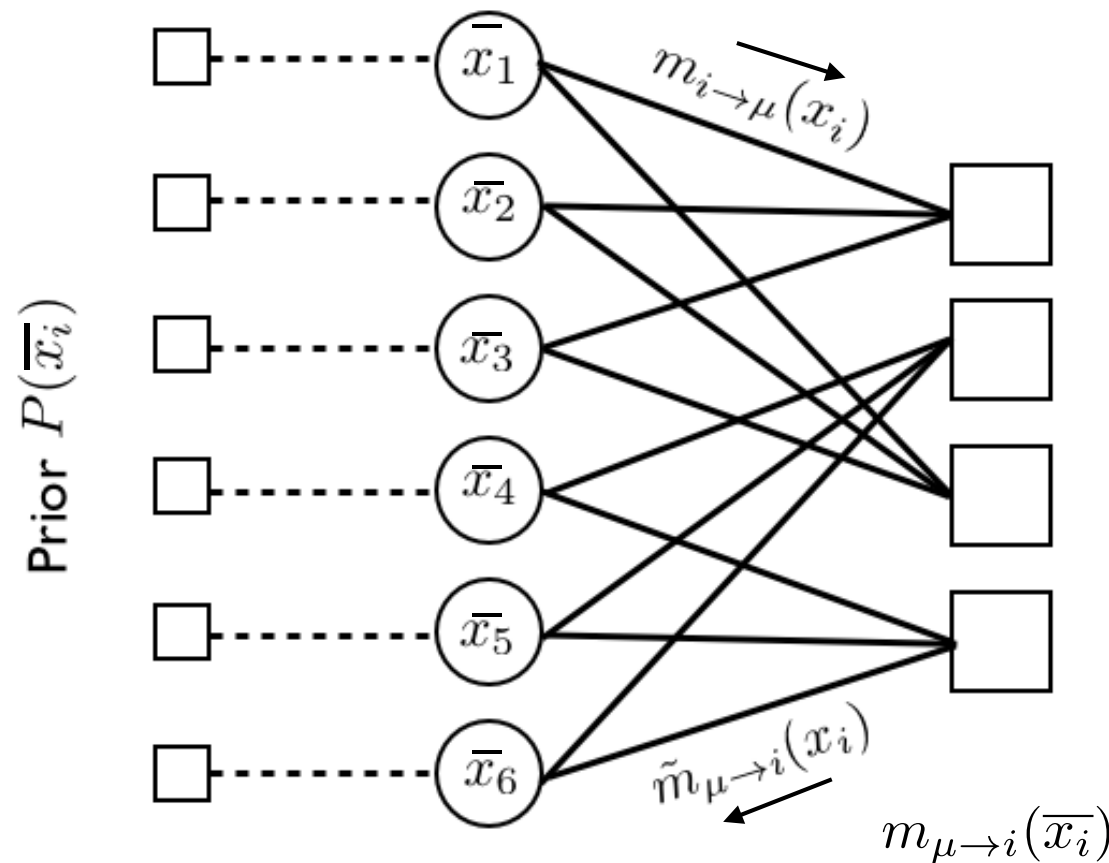
Belief Propagation?

Signal variables

Factor nodes

Not really an option ...

- i) Densely connected model
- ii) $O(L^2)$ messages
- iii) Combinatoric factor $O(B^L)$
- iv) Way too slow!



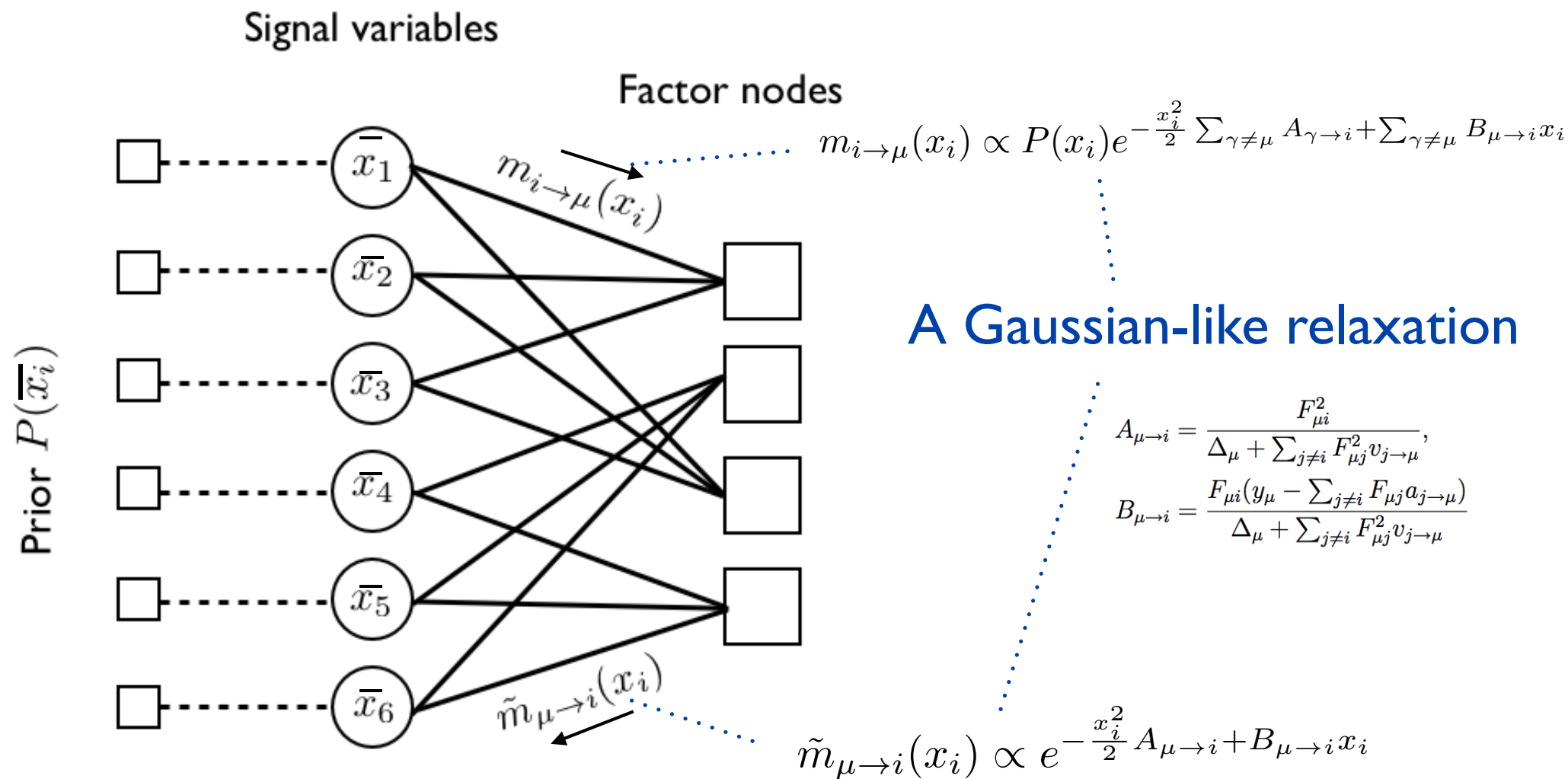
$$m_{i \rightarrow \mu}(\bar{x}_i) \propto P(\bar{x}_i) \prod_{\gamma \neq \mu} m_{\gamma \rightarrow i}(\bar{x}_i)$$

$$m_{\mu \rightarrow i}(\bar{x}_i) \propto \sum_{\{x_j\}} \prod_{j \neq i} m_{j \rightarrow \mu}(\bar{x}_i) e^{-\frac{SNR}{2} (y_\mu - \sum_l F_{\mu l} x_l)^2}$$

From belief propagation to AMP

Compute the marginal of $P(X|F, Y) \propto P(X)P(Y|F, X)$

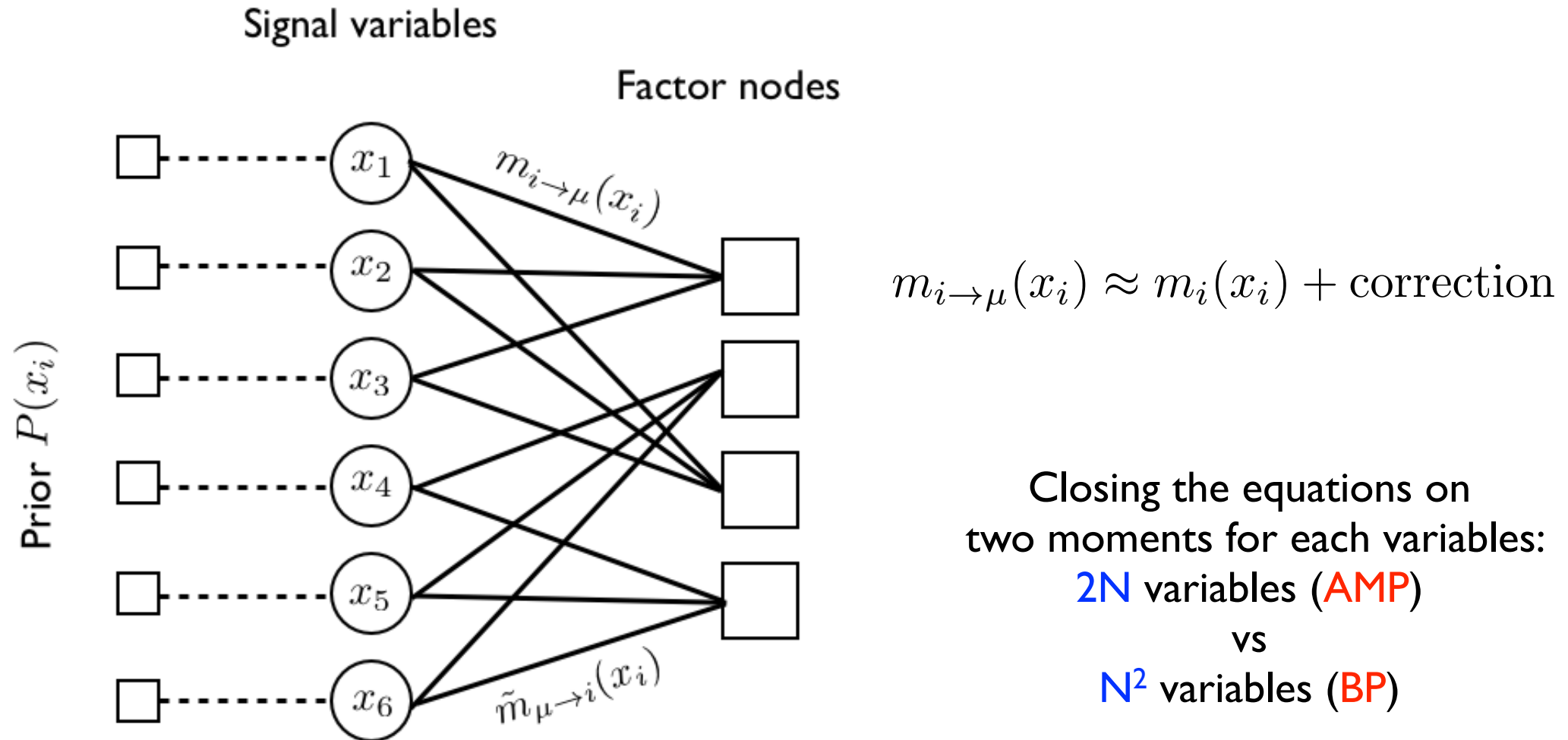
(i) A Gaussian relaxation (aka Gaussian-BP)



From belief propagation to AMP

Compute the marginal of $P(X|F, Y) \propto P(X)P(Y|F, X)$

- (i) A Gaussian relaxation (aka Gaussian-BP)
- (ii) Write the recursion in terms of single marginals (equivalent to “TAP” in Stat. Phys.)



From belief propagation to AMP

Compute the marginal of $P(X|F, Y) \propto P(X)P(Y|F, X)$

$$\begin{aligned}V_{\mu}^{t+1} &= \sum_i F_{\mu i}^2 v_i^t \\ \omega_{\mu}^{t+1} &= \sum_i F_{\mu i} a_i^t - (y_{\mu} - \omega_{\mu}^t) \frac{V_{\mu}^{t+1}}{1/\text{snr} + V_{\mu}^t} \\ (\Sigma_i^{t+1})^2 &= \left[\sum_{\mu} \frac{F_{\mu i}^2}{1/\text{snr} + V_{\mu}^{t+1}} \right]^{-1} \\ R_i^{t+1} &= a_i^t + (\Sigma_i^{t+1})^2 \sum_{\mu} F_{\mu i} \frac{(y_{\mu} - \omega_{\mu}^{t+1})}{1/\text{snr} + V_{\mu}^{t+1}} \\ a_i^{t+1} &= f_{a_i}(\{\Sigma_j^{t+1}, R_j^{t+1}\}_{j \in I}) \\ v_i^{t+1} &= f_{c_i}(\{\Sigma_j^{t+1}, R_j^{t+1}\}_{j \in I})\end{aligned}$$

$$\begin{aligned}a_i^t &:= f_{a_i}(\{\Sigma_j^t, R_j^t\}_{j \in I}) = \frac{e^{-\frac{1-2R_i^t}{2(\Sigma_i^t)^2}}}{\sum_{\{j \in I\}}^B e^{-\frac{1-2R_j^t}{2(\Sigma_j^t)^2}}} \\ v_i^t &:= f_{c_i}(\{\Sigma_j^t, R_j^t\}_{j \in I}) = a_i^t(1 - a_i^t)\end{aligned}$$

From belief propagation to AMP

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A (perhaps) more familiar form
from monday's talk
(Donoho, Montanari, Maleki '09)

$$\begin{aligned}a_{t+1} &= f_a^t(F' z_t + a_t) \\ z_t &= y - F a_t + \frac{1}{\alpha} \langle f_a^{t+1}(F' z^{t-1} + a_{t-1}) \rangle z^{t-1}\end{aligned}$$

Complexity is N^2 for iid matrix
 $N \log(N)$ for Hadamard/Fourier

Matlab implementation + demo: https://github.com/jeanbarbier/BPCS_common

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Asymptotic analysis

A heuristic computation of the optimal SER

Bayes-optimal estimate:
Marginalize w.r.t.

$$P(X|Y, F) = \frac{P(X)}{Z} \prod_{\mu=0}^M e^{-\frac{\text{SNR}}{2} (y_{\mu} - \sum_{i=1}^N F_{\mu i} \hat{x}_i)^2}$$

The replica method :Asymptotic estimates when $N \rightarrow \infty$
(statistical physics, information theory, optimization, etc..)

In CDMA

Tanaka '02

Guo,Verdu '06

In Compressed sensing

Rangan et al. '09

Kabashima '09

Guo Baron Shamai '09

Krzakala et al. '11

A heuristic computation of the optimal SER

Bayes-optimal estimate:
Marginalize w.r.t.

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The replica method : Asymptotic estimates when $N \rightarrow \infty$
(statistical physics, information theory, optimization, etc..)

1) Assume that $\log(Z)$ is concentrated

NON RIGOROUS

2) Use the following identity:

$$\overline{\log Z} = \lim_{n \rightarrow 0} \frac{\overline{Z^n} - 1}{n}$$

NON RIGOROUS

3) After (a bit) of work:

$$\overline{\log Z} \propto \int dE e^{N\Phi(E)} \longrightarrow MMSE = \max_E \Phi(E)$$

4) From the MSE E , one can compute the average SER

Single letter characterization of the SER

B-dimensional integral

$$\Phi_B(E) = -\frac{\log_2(B)}{2R} \left(\log(1/\text{snr} + E) + \frac{1-E}{1/\text{snr} + E} \right) + \int \mathcal{D}\bar{z} \log \left(e^{\frac{1}{2\Sigma(E)^2} + \frac{z_1}{\Sigma(E)}} + \sum_{i=2}^B e^{-\frac{1}{2\Sigma(E)^2} + \frac{z_i}{\Sigma(E)}} \right)$$

$$\text{with } \Sigma^t = \sqrt{\left(\frac{1}{\text{snr}} + E\right) \frac{R}{\log_2 B}} \quad \text{and} \quad \mathcal{D}\bar{z} = \prod dz_1 \frac{e^{-\frac{z_1^2}{2}}}{\sqrt{2\pi}} \dots dz_B \frac{e^{-\frac{z_B^2}{2}}}{\sqrt{2\pi}}$$

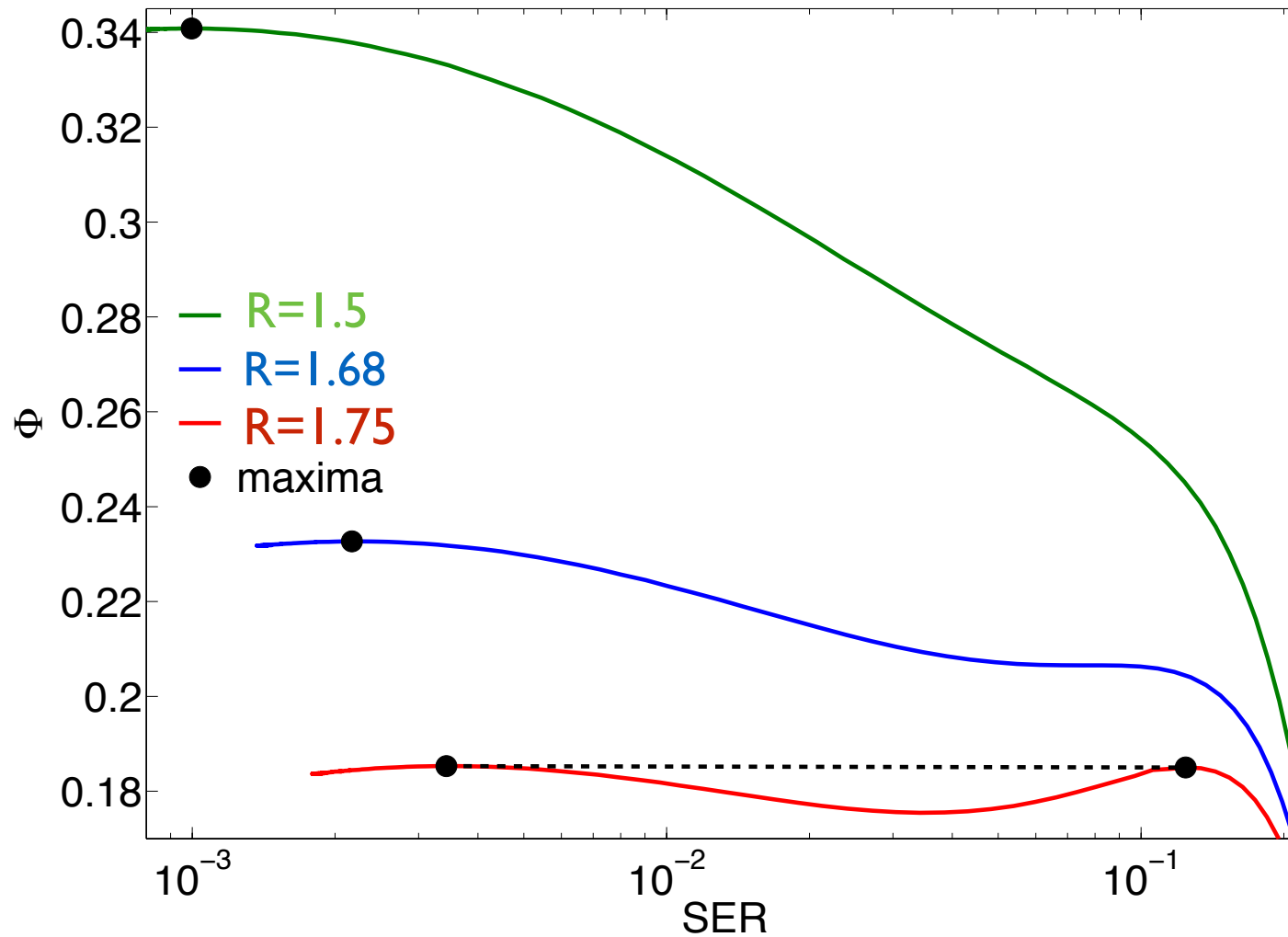
The SER can be computed from E as

$$\text{SER}^t = \int \mathcal{D}\bar{z} \mathbb{I} \left(\exists j \in \{2, \dots, B\} : f_{a_{j,1}}^{(0)}(\Sigma^t, \bar{z}) > f_{a_1}^{(1)}(\Sigma^t, \bar{z}) \right)$$

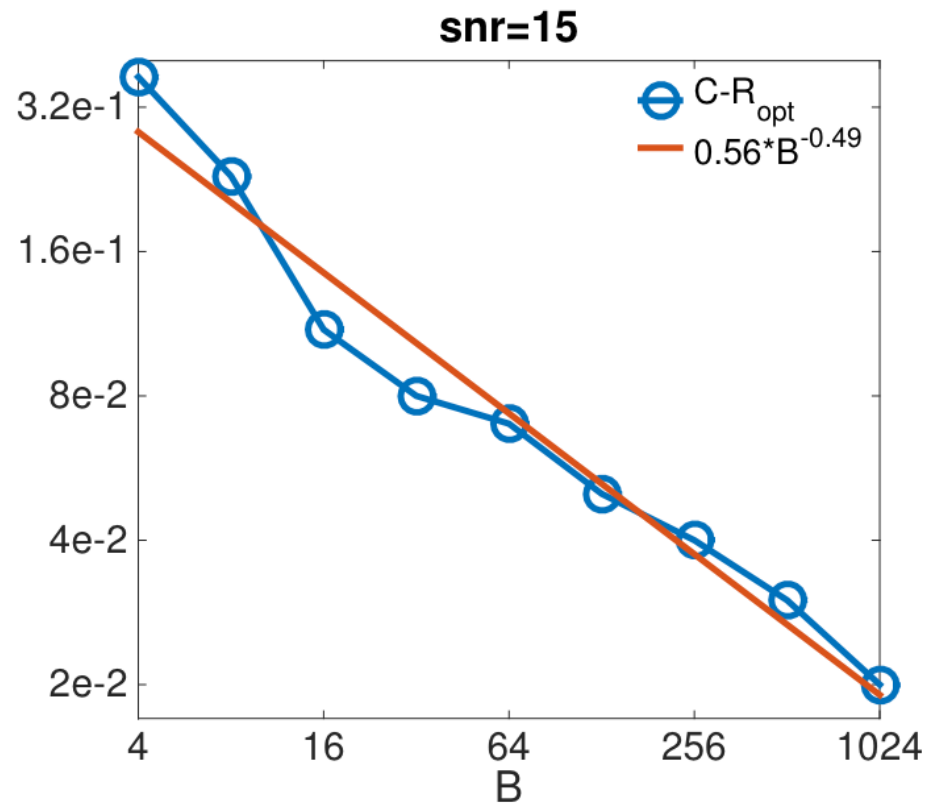
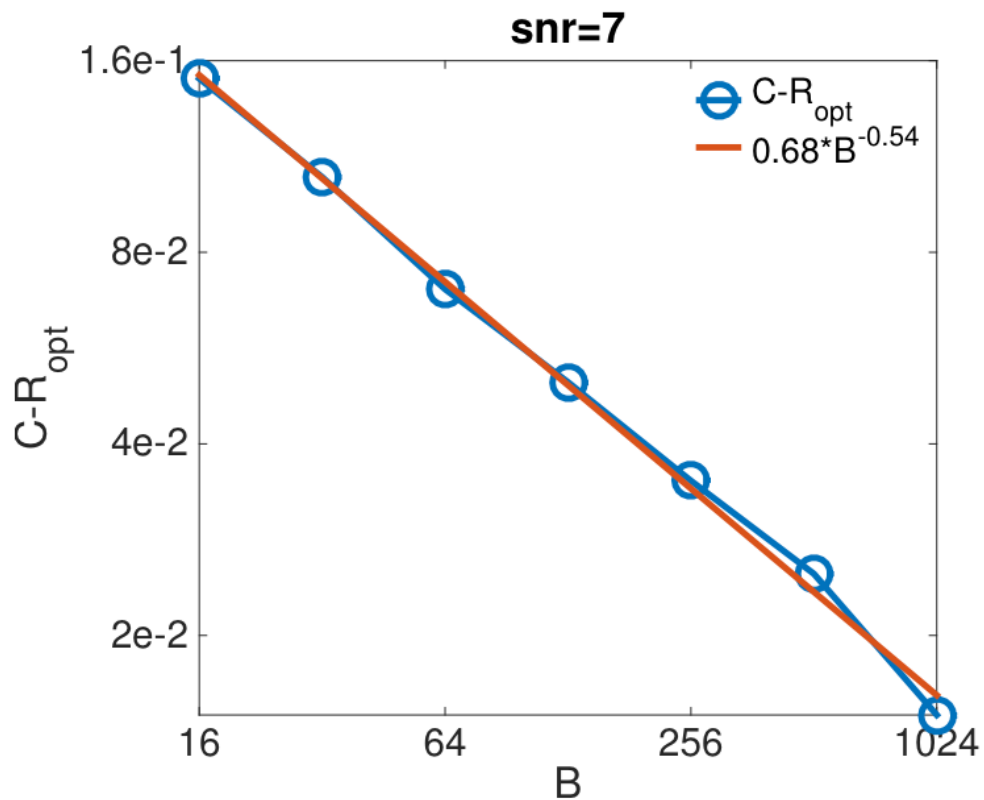
$$\text{with } \begin{cases} f_{a_i}^{(1)}(\Sigma, \bar{z}) = \left[1 + e^{-\frac{1}{\Sigma^2}} \sum_{\{1 \leq j \leq B: j \neq i\}} e^{\frac{z_j - z_i}{\Sigma}} \right]^{-1} \\ f_{a_{i,j}}^{(0)}(\Sigma, \bar{z}) = \left[1 + e^{\frac{1}{\Sigma^2} + \frac{z_j - z_i}{\Sigma}} + \sum_{\{1 \leq k \leq B: k \neq i, j\}} e^{\frac{z_k - z_i}{\Sigma}} \right]^{-1} \end{cases}$$

Single letter characterization of the SER

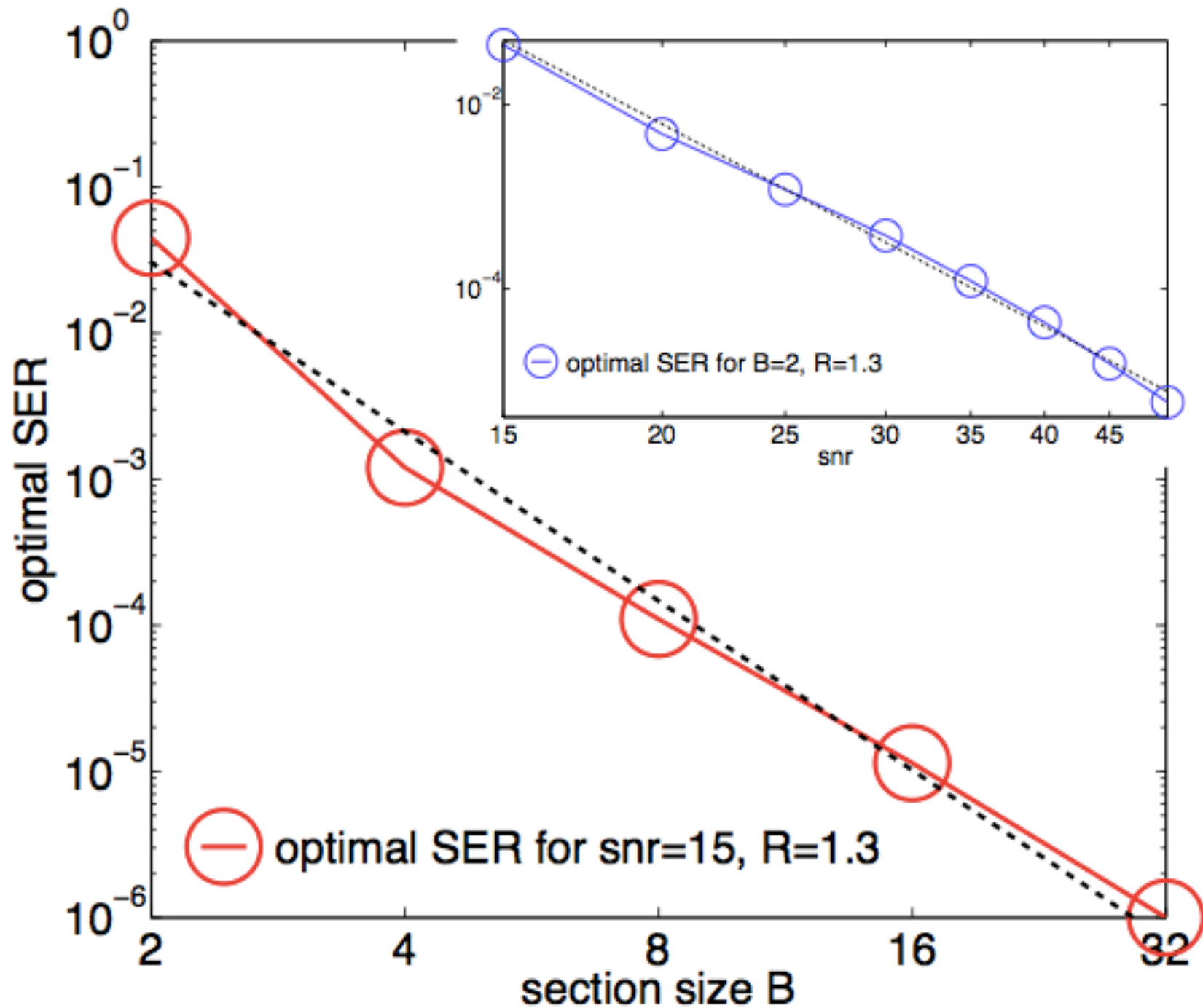
Here $B=2$, $\text{SNR}=15$, $C=2$



Gap to capacity closing polynomially



Error floor decay polynomially with B



State Evolution Analysis of AMP

(As in Bayati-Montanari '10 in the scalar case)

The MSE obey the following recursion:

$$E^t = \int \mathcal{D}\bar{z} \left([f_{a_1}^{(1)}(\Sigma^{t-1}, \bar{z}) - 1]^2 + (B-1)[f_{a_{1,2}}^{(0)}(\Sigma^{t-1}, \bar{z})]^2 \right)$$

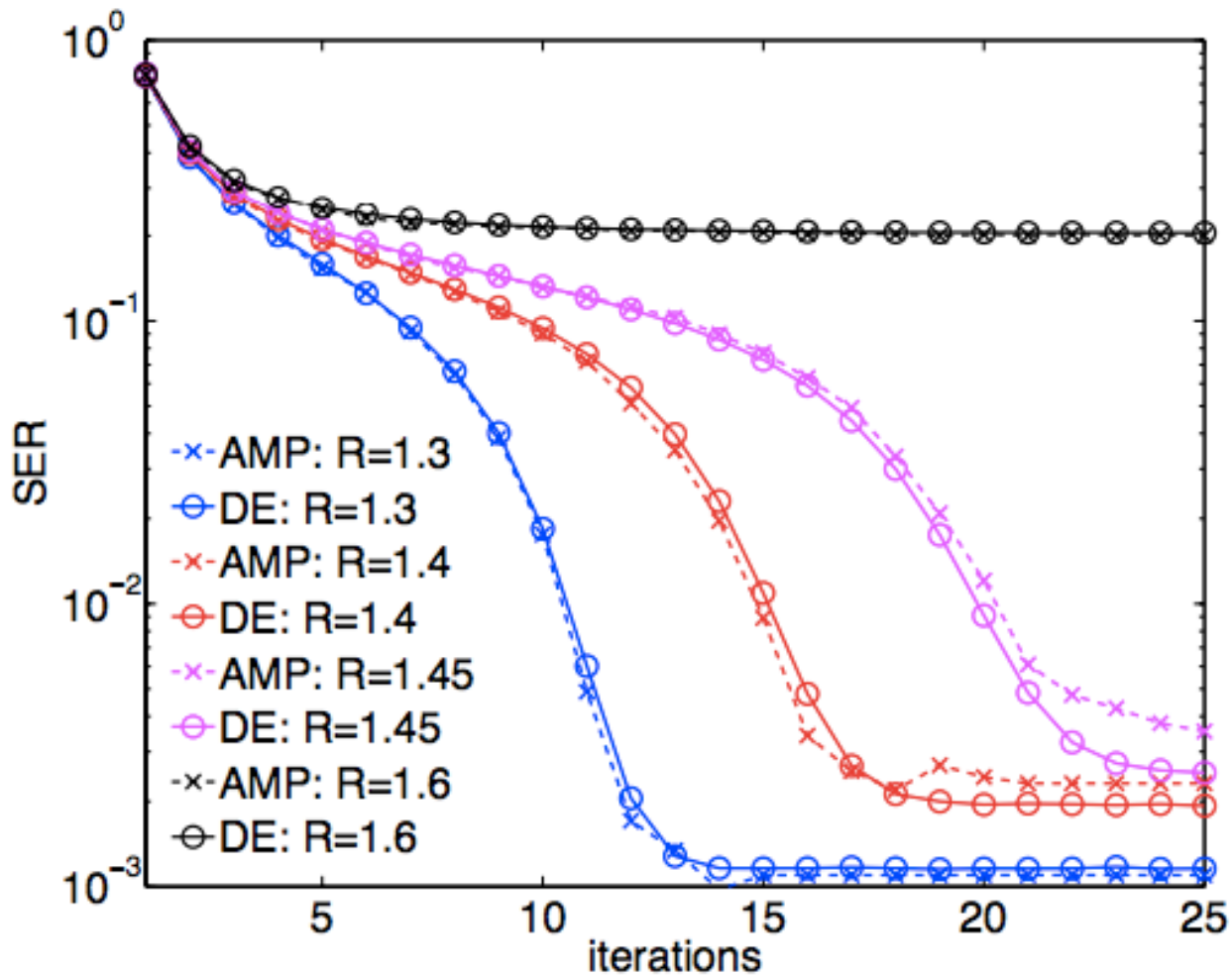
with: $f_{a_i}^{(1)}(\Sigma, \bar{z}) := \left[1 + e^{-\frac{1}{\Sigma^2}} \sum_{\{1 \leq j \leq B: j \neq i\}} e^{\frac{z_j - z_i}{\Sigma}} \right]^{-1}$ $f_{a_{i,j}}^{(0)}(\Sigma, \bar{z}) := \left[1 + e^{\frac{1}{\Sigma^2} + \frac{z_j - z_i}{\Sigma}} + \sum_{\{1 \leq k \leq B: k \neq i, j\}} e^{\frac{z_k - z_i}{\Sigma}} \right]^{-1}$

with $\Sigma^t := \sqrt{(1/\text{snr} + E^t)R/\log_2(B)}$, $E^t := \frac{1}{L} \sum_{i=1}^N (a_i^t - x_i)^2$

The Section Error Rate can be deduced from the MSE at all times by

$$\text{SER}^t = \int \mathcal{D}\bar{z} \mathbb{I} \left(\exists j \in \{2, \dots, B\} : f_{a_{j,1}}^{(0)}(\Sigma^t, \bar{z}) > f_{a_1}^{(1)}(\Sigma^t, \bar{z}) \right)$$

State Evolution vs AMP for finite sizes



$B=4$, $\text{SNR}=15$, $N=32768$, $R_{BP}=1.55$

State Evolution Analysis and the replica potential

$$E^t = \int \mathcal{D}\bar{z} \left([f_{a_1}^{(1)}(\Sigma^{t-1}, \bar{z}) - 1]^2 + (B-1)[f_{a_{1,2}}^{(0)}(\Sigma^{t-1}, \bar{z})]^2 \right)$$

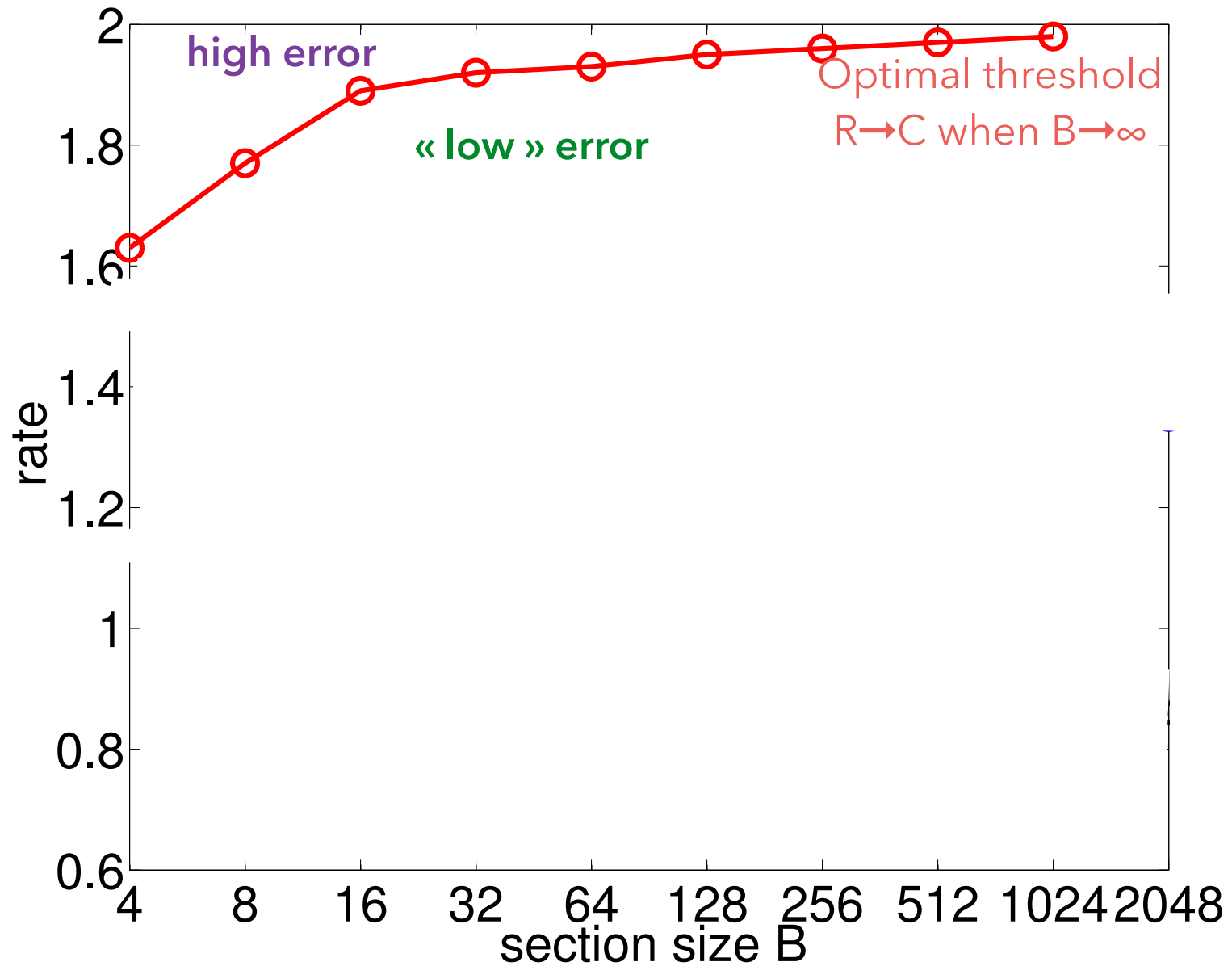
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Replica Potential:

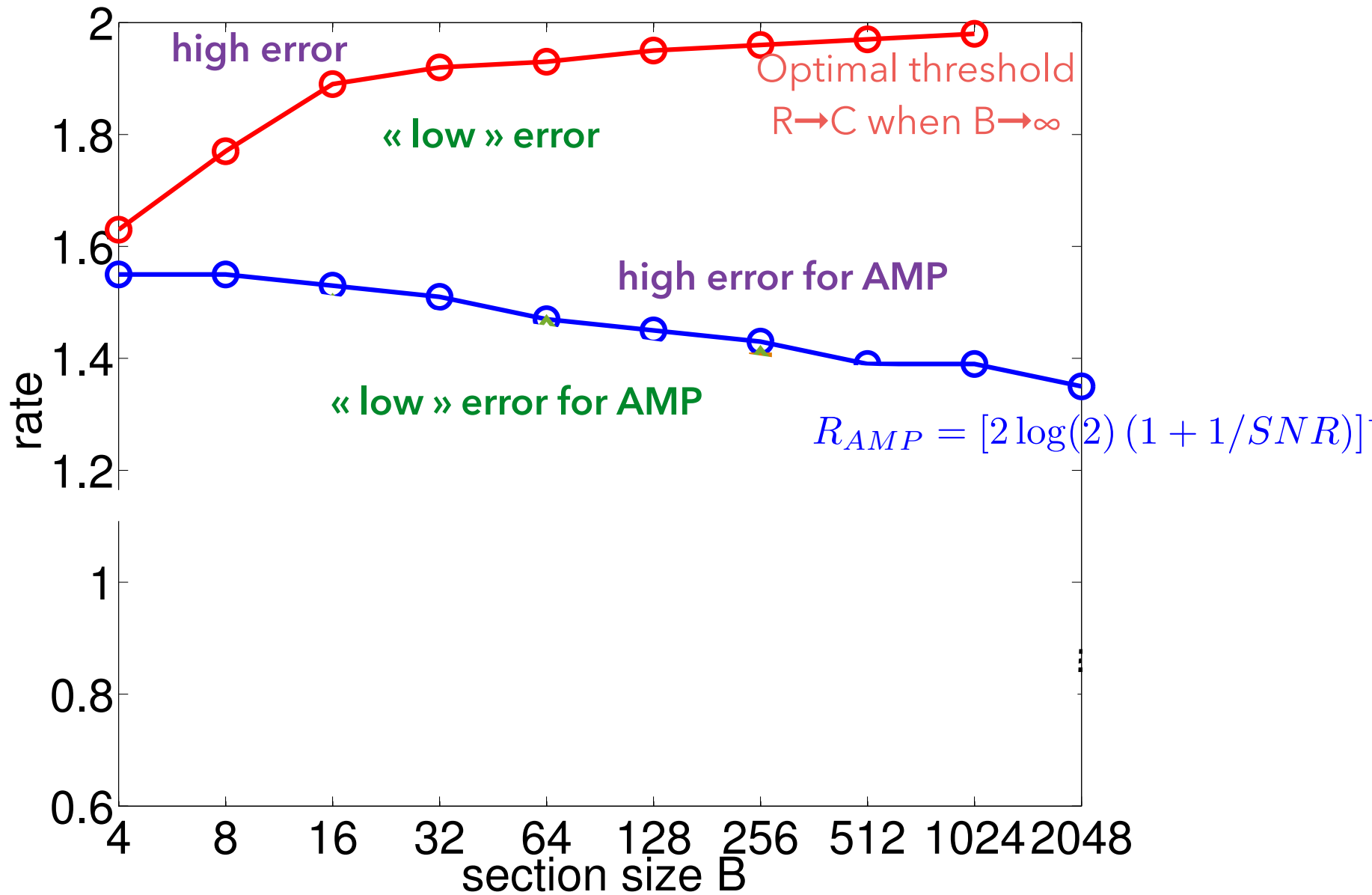
$$\Phi_B(E) = -\frac{\log_2(B)}{2R} \left(\log(1/\text{snr} + E) + \frac{1-E}{1/\text{snr} + E} \right) + \int \mathcal{D}\bar{z} \log \left(e^{\frac{1}{2\Sigma(E)^2} + \frac{z_1}{\Sigma(E)}} + \sum_{i=2}^B e^{-\frac{1}{2\Sigma(E)^2} + \frac{z_i}{\Sigma(E)}} \right)$$

The fixed point of the state evolution corresponds = extrema of the potential
Potential = Bethe free energy

Phase diagram $\text{SNR}=15$ ($C=2$)

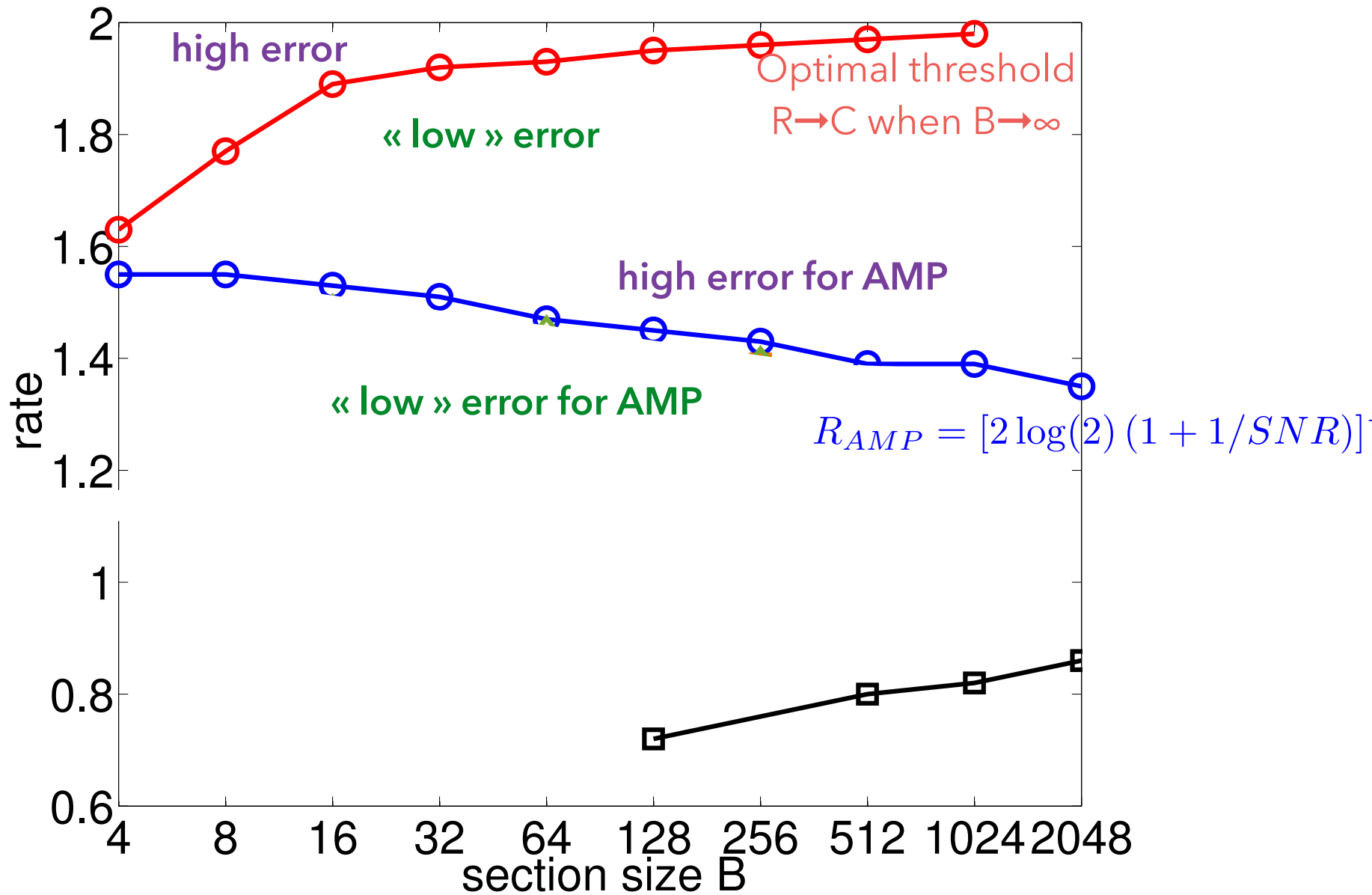


Phase diagram **SNR=15 (C=2)**



AMP gets worst as B increases :-)

Phase diagram **SNR=15 (C=2)**



... but is still way better than the alternative :-)

3

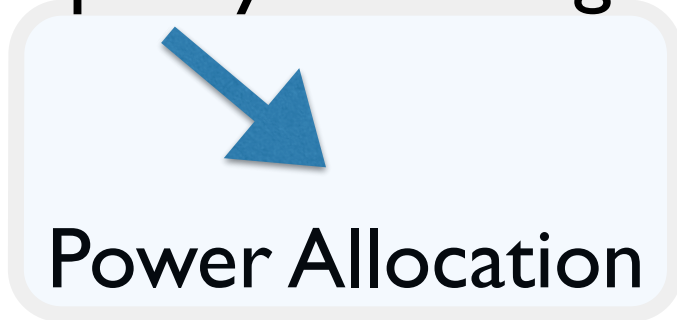
Two strategies to reach capacity

Two ways to make AMP capacity achieving:



Power Allocation

Two ways to make AMP capacity achieving:



vector $X =$

0		0	0		0	0	0		0	0	0			0	0	0
---	--	---	---	--	---	---	---	--	---	---	---	--	--	---	---	---



vector $X =$

0	c_1	0	0		c_2	0	0	0		0	0	0	c_3		c_4	0	0	0
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Generalization the State evolution for structured matrices shows that bAMP can reach capacity as $B \rightarrow \infty$

$$c_i^2 \propto 2^{-2Ci/L}$$

Power constraint: $\langle c_i^2 \rangle = 1$

In practice, however, this turns out to be not so efficient...

Two ways to make AMP capacity achieving:


Spatial coupling

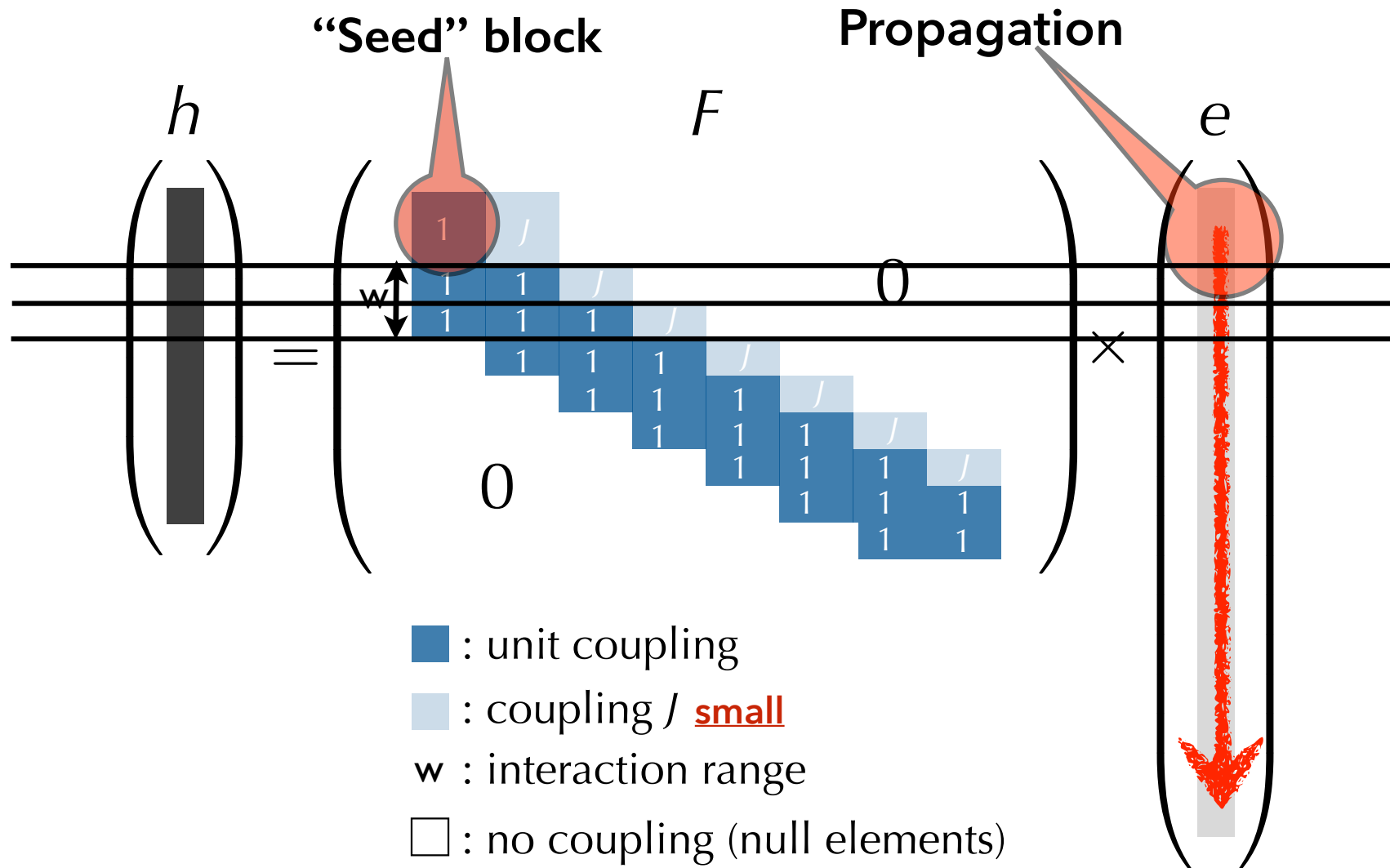
Use of a structured matrix

(Similar constructions in compressed sensing):

Pfister-Kuddekar '09

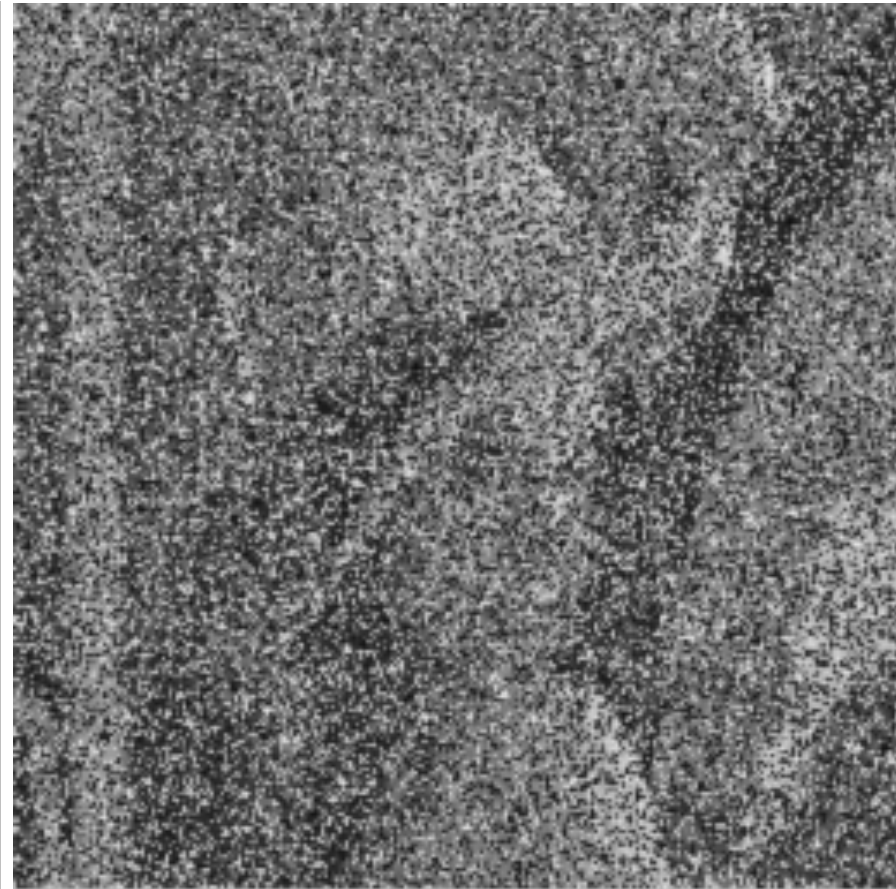
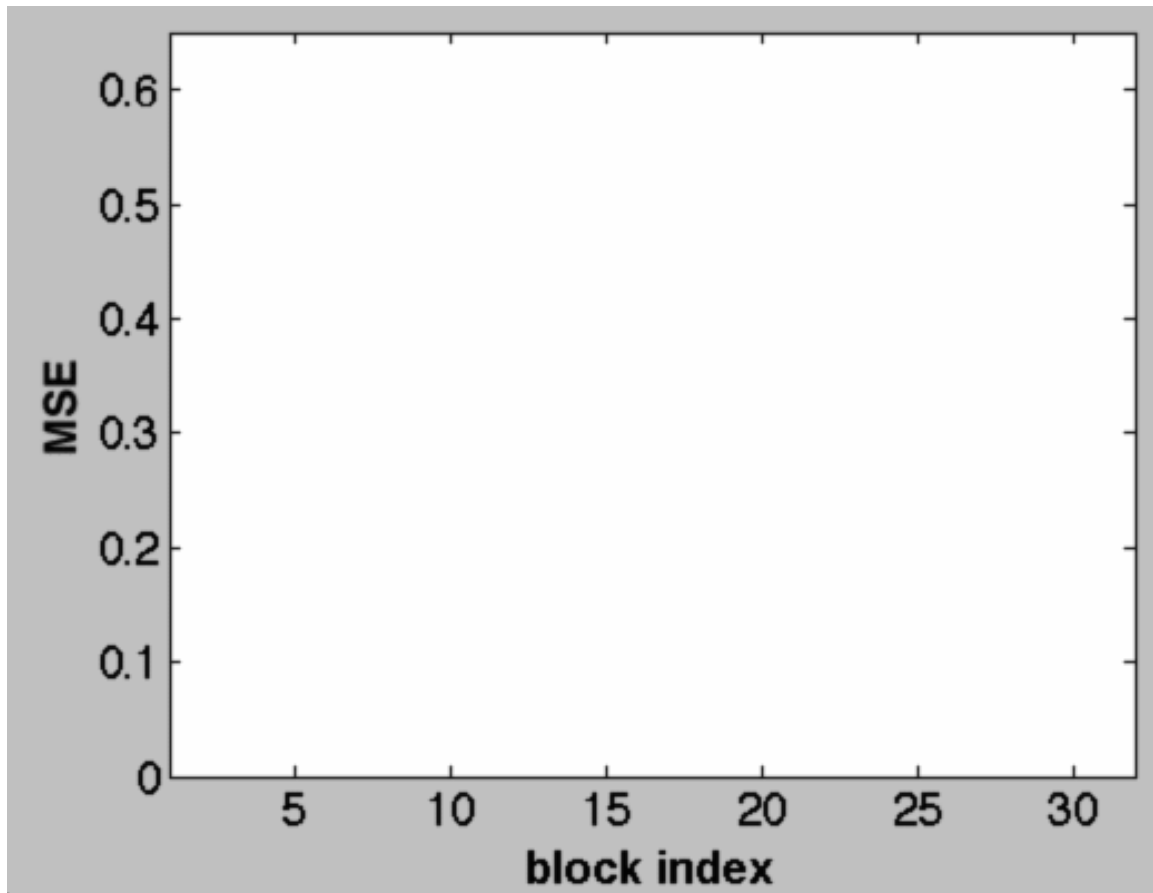
Krzakala et al '11

Donoho et al '11



Spatial coupling

Lena: $L=256^2$
B=256 levels of grey

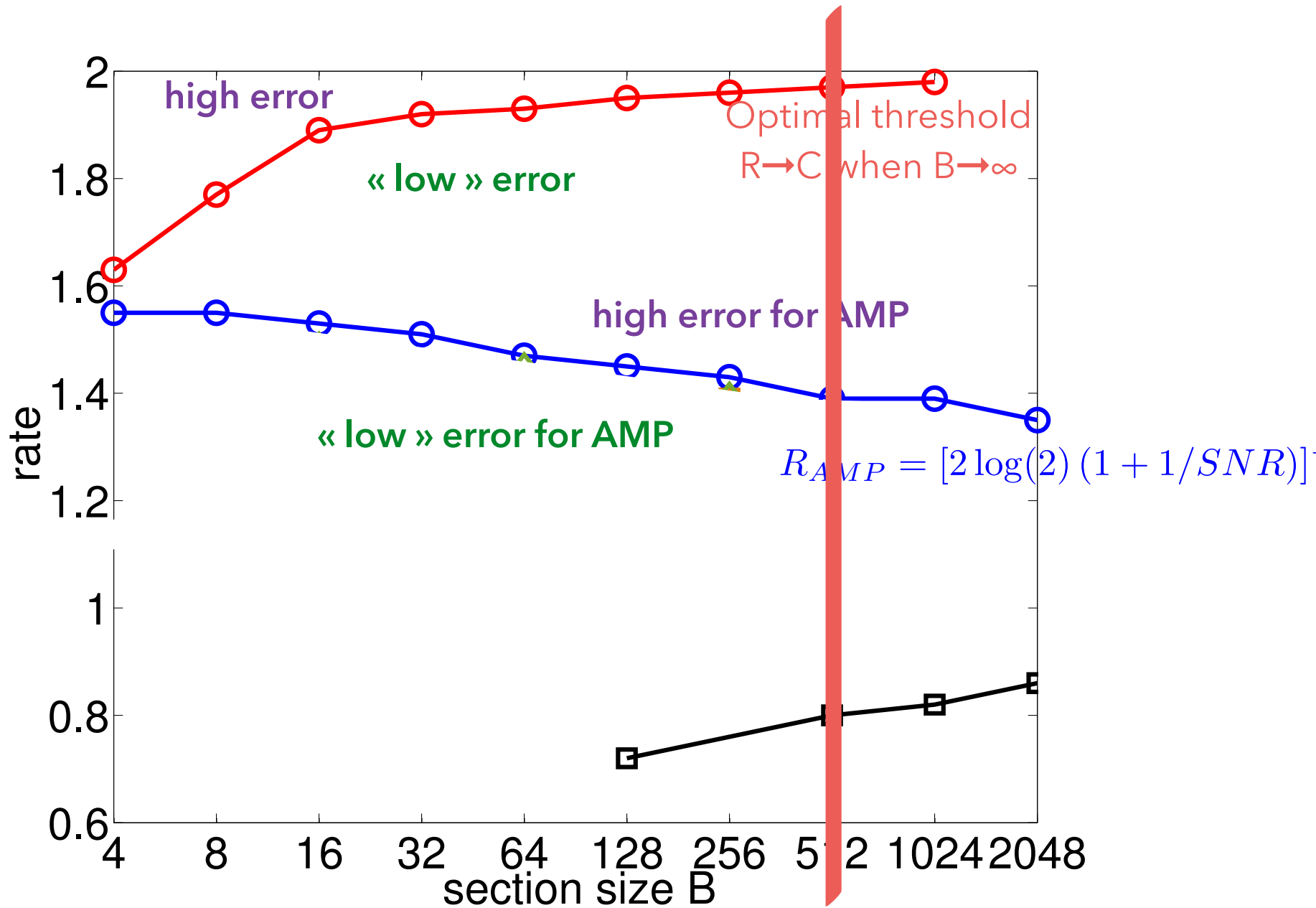


Matlab implementation + demo: https://github.com/jeanbarbier/BPCS_common



Practical tests!

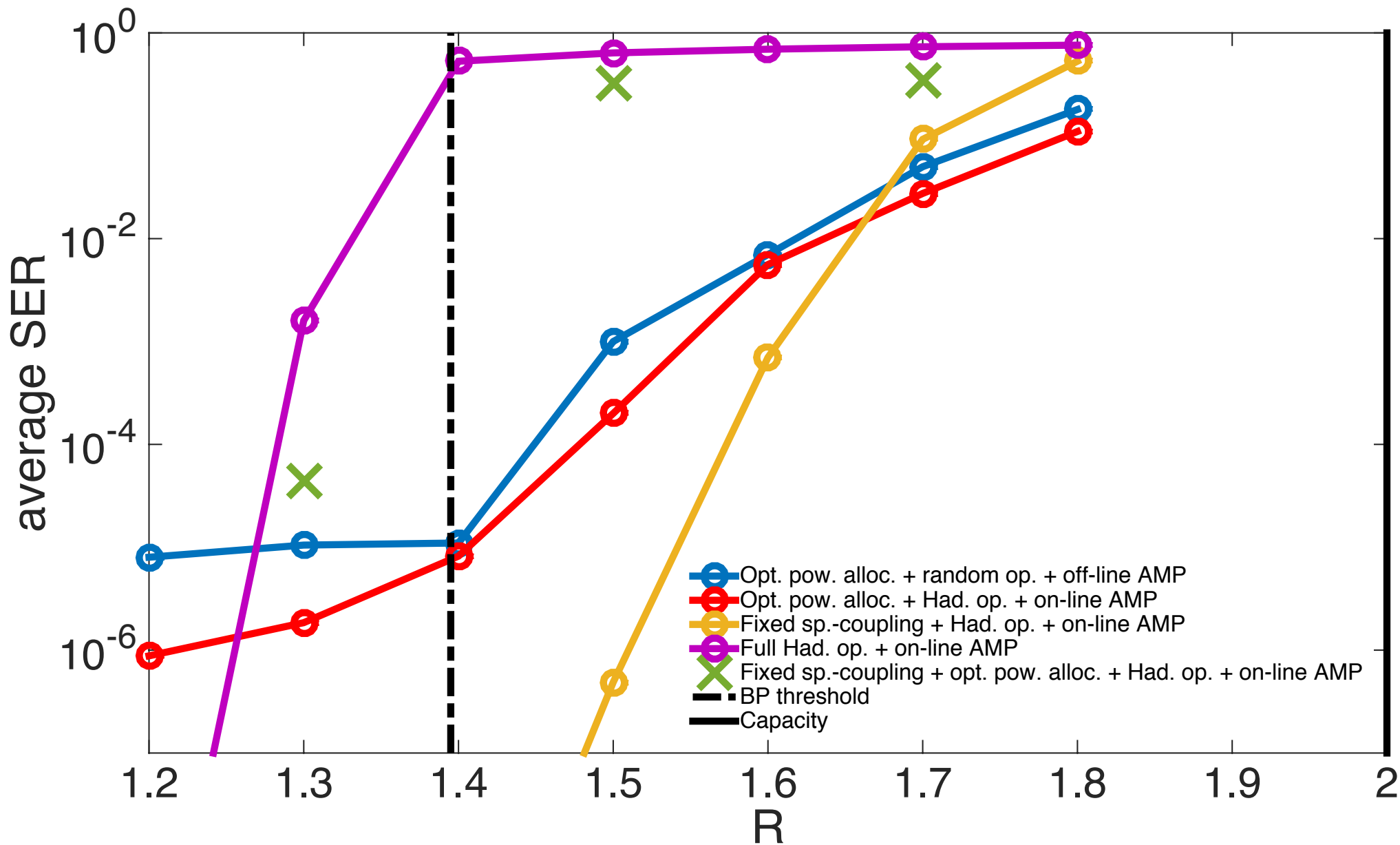
Phase diagram **SNR=15 (C=2)**

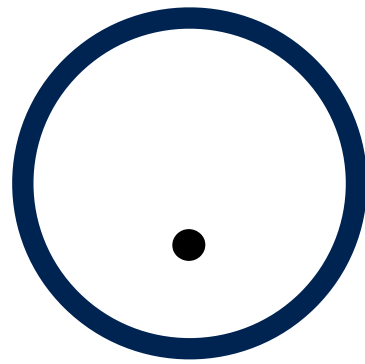


... but is still way better than the alternative :-)

Comparing everything
snr = 15, B = 512, L = 1024

Blocklength ~ 5000





Conclusion

- **Simple** coding, **capacity achieving** with spatial coupling/power allocation
- Similar phenomenology as LDPC codes
- **Efficient when used with structured operator** such as Hadamard
- Can be **analyzed** by state evolution/replica method
- Interesting **finite size** performances ?